

Parameterization of All Controllers for 4-block H^∞ Control Problems with Infinite and Finite $j\omega$ -Axis Zeros

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Abstract

A parameterization is presented of all solutions to the 4-block H^∞ control problem with infinite and finite $j\omega$ -axis zeros. The parameterization is given in terms of linear fractional transformation (LFT) on stable transfer function matrices with gain less than 1 which are free apart from satisfying certain interpolation conditions.

1 Introduction

Consider a generalized plant described as

$$\begin{aligned} \begin{bmatrix} z \\ y \end{bmatrix} &= P(s) \begin{bmatrix} w \\ u \end{bmatrix} = \begin{bmatrix} P_{11} & P_{12} \\ P_{21} & P_{22} \end{bmatrix} \begin{bmatrix} w \\ u \end{bmatrix} \\ &= \left[\begin{array}{c|cc} A & B_1 & B_2 \\ \hline C_1 & 0 & D_{12} \\ C_2 & D_{21} & 0 \end{array} \right] \begin{bmatrix} w \\ u \end{bmatrix}, \end{aligned} \quad (1)$$

where $z \in R^m$, $y \in R^q$, $w \in R^r$ and $u \in R^p$ are the controlled error, the observation output, the exogenous input and the control input, respectively. The H^∞ control problem is to find a *proper* control law $u(s) = K(s)y(s)$ which internally stabilizes the closed-loop system and satisfies $\|\Phi(s)\|_\infty < 1$, where $\Phi(s)$ is the closed-loop transfer function from w to z given by the following LFT:

$$\Phi(s) = F_l(P; K) := P_{11} + P_{12}K(I - P_{22}K)^{-1}P_{21}. \quad (2)$$

It is well known that the *standard* H^∞ control problems has been solved when plant (1) satisfies the following assumptions:

- (A₁) (A, B_2, C_2) is stabilizable and detectable.
- (A₂) $\text{rank } D_{12} = p$, $\text{rank } D_{21} = q$.
- (A₃) $P_{12}(s)$ and $P_{21}(s)$ have no $j\omega$ -axis invariant zeros.

In this paper, instead of assumptions (A₂) and (A₃), we assume that

(A₄) $P_{12}(s)$ and $P_{21}(s)$ have full normal column and row ranks, respectively.

As to the above H^∞ control problem, via the eigenstructures related to infinite and finite $j\omega$ -zeros, [1] presented the necessary and sufficient conditions for checking its solvability by solving two reduced-order Riccati equations and examining matrix norm conditions related to $j\omega$ -axis zeros. Thus the above paper extended the DGKF's approach [2] to H^∞ control problems without the constraints on the infinite or finite $j\omega$ -axis zeros. As a sequel of [1], we discuss the parameterization of all H^∞ controllers in this paper.

Notations: The open left and right half complex plane are denoted by C_- and C_+ , respectively. The $j\omega$ -axis and $j\omega$ -axis with infinity are denoted by Ω and Ω_e , respectively. The set of all $m \times r$ constant real matrices is denoted by $R^{m \times r}$. I_r denotes the identity matrix of size $r \times r$. $RH_{m \times r}^\infty$ denotes the set of all $m \times r$ rational stable proper matrices, and $BH_{m \times r}^\infty$ denotes the subset of $RH_{m \times r}^\infty$ with H^∞ -norm less than 1. $\sigma(A)$ denotes the set of all eigenvalues of matrix A . $\rho(X)$ is the maximum eigenvalue of X . We express the star product of M_1 and M_2 by $M = M_1 * M_2$ so that $F_l(M_1, F_l(M_2, K)) = F_l(M_1 * M_2, K)$ holds.

2 Preliminaries

2.1 Infinite eigenstructures

Denote the system matrix pencils of $P_{12}(s)$ and $P_{21}^T(s)$ as $-sP_E + P_A$ and $-s\tilde{P}_E + \tilde{P}_A$, respectively, where

$$P_E := \begin{bmatrix} I_n & 0 \\ 0 & 0 \end{bmatrix}, \quad P_A := \begin{bmatrix} A & B_2 \\ C_1 & D_{12} \end{bmatrix}. \quad (3)$$

$$\tilde{P}_E := \begin{bmatrix} I_n & 0 \\ 0 & 0 \end{bmatrix}, \quad \tilde{P}_A := \begin{bmatrix} A^T & C_2^T \\ B_1^T & D_{21}^T \end{bmatrix}. \quad (4)$$

According to assumption (A₄), the above two pencils has full normal column ranks.

Let (v_1^1, \dots, v_p^1) be a base of $\text{Ker } P_E$. Then the infinite eigenvectors are defined by

$$P_E v_j^1 = 0, \quad j = 1, \dots, p, \quad (5)$$

$$P_E v_j^{k+1} = P_A v_j^k, \quad k = 1, \dots, k_j - 1, \quad (6)$$

where $v_j^{k_j}$ is the last one of each infinite eigenvector chain satisfying $P_A v_j^{k_j} \notin \text{Im } P_E$. Now construct

$$V_\infty := [V_r \quad V_h], \quad (7)$$

where $V_h \in R^{(n+p) \times p}$ contains all the *last* infinite eigenvectors and V_r are the remainders. Therefore, the complete infinite eigenstructure of $-sP_E + P_A$ is defined by

$$(-sP_E + P_A)V_\infty = P_A V_\infty (-sN + I), \quad (8)$$

where N is a nilpotent matrix. From (6), we know that $[C_1 \quad D_{12}] V_r = 0$, then decompose

$$P_A V_\infty = \begin{bmatrix} A & B_2 \\ C_1 & D_{12} \end{bmatrix} [V_r \quad V_h] =: \begin{bmatrix} T & \hat{B}_2 \\ 0 & \hat{D}_{12} \end{bmatrix}, \quad (9)$$

which yields

$$T := [A \quad B_2] V_r, \quad \hat{B} := [A \quad B_2] V_h, \quad (10)$$

$$\hat{D} := [C_1 \quad D_{12}] V_h. \quad (11)$$

Note that \hat{D}_{12} is *injective* [5].

Dually consider $P_{21}^T(-s)$. Now arrange all the infinite eigenvectors of $-s\tilde{P}_E + \tilde{P}_A$ as

$$\tilde{V}_\infty := [\tilde{V}_r \quad \tilde{V}_h], \quad (12)$$

where $\tilde{V}_h \in R^{(n+q) \times q}$ contains all the *last* infinite eigenvectors and \tilde{V}_r are the reminders. From $\tilde{P}_A \tilde{V}_\infty$, define

$$\tilde{T} := [A^T \quad C_2^T] \tilde{V}_r, \quad \hat{C}_2^T := [A^T \quad C_2^T] \tilde{V}_h, \quad (13)$$

$$\hat{D}_{21}^T := [B_1^T \quad D_{21}^T] \tilde{V}_h, \quad (14)$$

which follows that \hat{D}_{21}^T is also *injective*.

2.2 Finite $j\omega$ -axis eigenstructures

Let the $j\omega$ -axis eigenspaces of $-sP_E + P_A$ and $-s\tilde{P}_E + \tilde{P}_A$ be spanned by real $\begin{bmatrix} T_1 \\ T_2 \end{bmatrix}$ and $\begin{bmatrix} \tilde{T}_1 \\ \tilde{T}_2 \end{bmatrix}$, respectively.

It follows that there exist Λ_j and $\tilde{\Lambda}_j$ such that $\sigma(\Lambda_j) \subset \Omega$ and $\sigma(\tilde{\Lambda}_j) \subset \Omega$ hold, and

$$(-sP_E + P_A) \begin{bmatrix} T_1 \\ T_2 \end{bmatrix} = \begin{bmatrix} T_1 \\ 0 \end{bmatrix} (-sI + \Lambda_j), \quad (15)$$

$$(-s\tilde{P}_E + \tilde{P}_A) \begin{bmatrix} \tilde{T}_1 \\ \tilde{T}_2 \end{bmatrix} = \begin{bmatrix} \tilde{T}_1 \\ 0 \end{bmatrix} (-sI + \tilde{\Lambda}_j). \quad (16)$$

2.3 Stable eigenstructures

Denote the system matrices of $P_{12}^T(-s)P_{12}(s)$ and $P_{21}(-s)P_{21}^T(s)$ as:

$$W_{12}(s) := \begin{bmatrix} -sI + A & 0 & B_2 \\ -C_1^T C_1 & -sI - A^T & -C_1^T D_{12} \\ D_{12}^T C_1 & B_2^T & D_{12}^T D_{12} \end{bmatrix}, \quad (17)$$

$$W_{21}(s) := \begin{bmatrix} -sI + A^T & 0 & C_2^T \\ -B_1 B_1^T & -sI - A & -B_1 D_{21}^T \\ D_{21} B_1^T & C_2 & D_{21} D_{21}^T \end{bmatrix}. \quad (18)$$

Let $\begin{bmatrix} U_1 \\ U_2 \\ U_3 \end{bmatrix}$ and $\begin{bmatrix} \tilde{U}_1 \\ \tilde{U}_2 \\ \tilde{U}_3 \end{bmatrix}$ spanned stable eigenspace of $W_{12}(s)$ and $W_{21}(s)$, respectively. There exist stable Λ_{12} and Λ_{21} such that

$$W_{12}(s) \begin{bmatrix} U_1 \\ U_2 \\ U_3 \end{bmatrix} = \begin{bmatrix} U_1 \\ U_2 \\ 0 \end{bmatrix} (-sI + \Lambda_{12}), \quad (19)$$

$$W_{21}(s) \begin{bmatrix} \tilde{U}_1 \\ \tilde{U}_2 \\ \tilde{U}_3 \end{bmatrix} = \begin{bmatrix} \tilde{U}_1 \\ \tilde{U}_2 \\ 0 \end{bmatrix} (-sI + \Lambda_{21}). \quad (20)$$

From [5], we can obtain

LEMMA 1 Under the assumptions (A₁) and (A₄). Then S and \tilde{S} are nonsingular, where

$$S := [U_1 \quad T_1 \quad T], \quad \tilde{S} := [\tilde{U}_1 \quad \tilde{T}_1 \quad \tilde{T}]. \quad (21)$$

2.4 Solvability Conditions

LEMMA 2 [1] Under the assumptions (A₁) and (A₄), the H^∞ control problem for plant $P(s)$ in (1) is solvable if and only if the following statement holds.

(i) The following Riccati equation has a stabilizing solution $X_r \geq 0$,

$$(A_r - \hat{B}_{r2} E_{12}^{-1} \hat{D}_{12}^T C_{r1})^T X_r + X_r (A_r - \hat{B}_{r2} E_{12}^{-1} \hat{D}_{12}^T C_{r1}) + X_r (B_{r1} B_{r1}^T - \hat{B}_{r2} E_{12}^{-1} \hat{B}_{r2}^T) X_r + C_{r1}^T (I - \hat{D}_{12} E_{12}^{-1} \hat{D}_{12}^T) C_{r1} = 0, \quad (22)$$

where $E_{12} := \hat{D}_{12}^T \hat{D}_{12}$, and

$$A_r := L_1 A U_1, \quad B_{r1} := L_1 B_1, \quad (23)$$

$$\hat{B}_{r2} := L_1 \hat{B}_2, \quad C_{r1} := C_1 U_1, \quad (24)$$

$$[L_1^T \quad L_2^T \quad L_3^T]^T := [U_1 \quad T_1 \quad T]^{-1}, \quad (25)$$

where T, \hat{B}_2 are given by (10), \hat{D}_{12} is given by (11), T_1 is given by (15), and U_1 is given by (19), respectively.

(ii) The following Riccati equation has a stabilizing solution $Y_r \geq 0$,

$$Y_r(\bar{A}_r - \bar{B}_{r1}\hat{D}_{21}^T E_{21}^{-1}\hat{C}_{r2})^T + (\bar{A}_r - \bar{B}_{r1}\hat{D}_{21}^T E_{21}^{-1}\hat{C}_{r2})Y_r + Y_r(\bar{C}_{r1}^T \bar{C}_{r1} - \hat{C}_{r2}^T E_{21}^{-1}\hat{C}_{r2})Y + \bar{B}_{r1}(I - \hat{D}_{21}^T E_{21}^{-1}\hat{D}_{21})\bar{B}_{r1}^T = 0, \quad (26)$$

where $E_{21} := \hat{D}_{21}\hat{D}_{21}^T$, and

$$\bar{A}_r := \bar{U}_1^T A \bar{L}_1^T, \quad \bar{B}_{r1} := \bar{U}_1^T B_1, \quad (27)$$

$$\hat{C}_{r2} := \hat{C}_2 \bar{L}_1^T, \quad \bar{C}_{r1} := \bar{U}_1^T C_1, \quad (28)$$

$$[\bar{L}_1^T \quad \bar{L}_2^T \quad \bar{L}_3^T]^T := [\bar{U}_1 \quad \bar{T}_1 \quad \bar{T}]^{-1}, \quad (29)$$

where \bar{T} , \hat{C}_2 are given by (13), \hat{D}_{21} is given by (14), \bar{T}_1 is given by (16), and \bar{U}_1 is given by (20), respectively.

(iii) $\rho(XY) < 1$, where

$$X := L_1^T X_r L_1, \quad Y := \bar{L}_1^T Y_r \bar{L}_1. \quad (30)$$

(iv)

$$U_{12i}^* U_{12i} > X_{12i}^* B_{n1} B_{n1}^T X_{12i}, \quad (31)$$

$$U_{21i}^* U_{21i} > X_{21i}^* C_{n1}^T C_{n1} X_{21i}, \quad (32)$$

where X_{12i} , U_{12i} , X_{21i} and U_{21i} satisfy

$$\begin{bmatrix} -j\omega_i I + A_n^T & C_{n1}^T \\ B_{n2}^T & N_{12}^T \end{bmatrix} \begin{bmatrix} X_{12i} \\ U_{12i} \end{bmatrix} = 0, \quad (33)$$

$$\begin{bmatrix} -j\omega_i I + A_n & B_{n1} \\ C_{n2} & N_{21} \end{bmatrix} \begin{bmatrix} X_{21i} \\ U_{21i} \end{bmatrix} = 0, \quad (34)$$

where A_n , B_{n1} , B_{n2} , C_{n1} , C_{n2} are defined by the following new plant as:

$$P_n(s) = \begin{bmatrix} P_{n11} & P_{n12} \\ P_{n21} & P_{n22} \end{bmatrix} = \begin{bmatrix} A_n & B_{n1} & B_{n2} \\ C_{n1} & 0 & N_{12} \\ C_{n2} & N_{21} & 0 \end{bmatrix} \quad (35)$$

with its matrices defined by

$$A_n := A + B_1 B_1^T X + ZY F_\infty^T F_\infty, \quad (36)$$

$$B_{n1} := -ZL_\infty, \quad B_{n2} := B_2 - ZY F_\infty^T N_{12}, \quad (37)$$

$$C_{n1} := -F_\infty, \quad C_{n2} := (C_2 - N_{21} L_\infty^T X Z) Z^{-1}, \quad (38)$$

$$Z := (I - YX)^{-1}, \quad (39)$$

$$F_\infty := -E_{12}^{-1/2}(\hat{B}_2^T X + \hat{D}_{12}^T C_1), \quad (40)$$

$$L_\infty := -(Y\hat{C}_2^T + B_1\hat{D}_{21}^T)E_{21}^{-1/2}, \quad (41)$$

$$N_{12} := E_{12}^{-1/2}\hat{D}_{12}^T D_{12}, \quad (42)$$

$$N_{21} := (E_{21}^{-1/2}\hat{D}_{21} D_{21}^T)^T. \quad (43)$$

LEMMA 3 Under the assumptions (A_1) and (A_4) , if Conditions (i)-(iv) in Theorem 2 hold, then

(i) $P_{n12}(s)$ has no invariant zeros in C_+ . Moreover, (A_n, B_{n2}) is stabilizable.

(ii) $P_{n21}(s)$ has no invariant zeros in C_+ . Moreover, and (A_n, C_{n2}) is detectable.

3 Parameterization of All H^∞ Controllers

From Lemmas 2 and 3, we have

THEOREM 1 Suppose that the H^∞ control problem is solvable. All H^∞ controllers are

$$K(s) = F_l(M^\infty(s), S(s)), \quad (44)$$

where $S(s) \in BH^\infty$ such that $K(s)$ is proper and stabilizes the closed-loop system, and

$$M^\infty(s) = J P_n^{-1}(s) J = \begin{bmatrix} A_n & B_{n2} & B_{n1} \\ C_{n2} & 0 & N_{21} \\ C_{n1} & N_{12} & 0 \end{bmatrix}^{-1}, \quad (45)$$

with

$$J = \begin{bmatrix} 0 & I \\ I & 0 \end{bmatrix}.$$

Proof Suppose that K_0 is an H^∞ controller. Then K_0 stabilizes the closed-loop such that

$$\Phi_0 = F_l(P_n, K_0) \in BH^\infty. \quad (46)$$

Take $S_0 = \Phi_0 \in BH^\infty$, it follows that

$$K_0 = F_l(M^\infty, S_0) \quad (47)$$

In other words, any solution can be expressed by (44).

On the other hand, let S_1 be an arbitrary BH^∞ matrix such that

$$K_1 = F_l(M^\infty, S_1) \quad (48)$$

is proper and stabilizes the closed-loop system. Then

$$\begin{aligned} \Phi_1 &= F_l(P_n, K_1) = F_l(P_n * M^\infty, S_1) \\ &= F_l\left(\begin{bmatrix} 0 & I \\ I & 0 \end{bmatrix}, S_1\right) = S_1 \in BH^\infty. \end{aligned} \quad (49)$$

Therefore, $K_1(s)$ is a desired H^∞ controller. \square

Remark 3.1 Under assumption (A_4) , then $P_{n12}(s)$ and $P_{n21}(s)$ has no invariant zeros in C_+ , which follows that $P_n^{-1}(s)$ exists. If assumption (A_2) holds, from (10), (11), (13) and (14), we can choose $V_\infty = V_h = \begin{bmatrix} 0 & I_p \end{bmatrix}^T$ and $\check{V}_\infty = \check{V}_h = \begin{bmatrix} 0 & I_q \end{bmatrix}^T$, we obtain $\hat{B}_2 = B_2$, $\hat{D}_{12} = D_{12}$, $\hat{C}_1 = C_1$, $\hat{D}_{21} = D_{21}$. Then $K(s)$ in (44) is proper for all $S(s) \in BH^\infty$. If both assumption (A_2) and assumption (A_3) hold, we can choose $L_1 = U_1 = \bar{L}_1 = \bar{U}_1 = I_n$, and for all $S(s) \in BH^\infty$, $K(s)$ in (44) are H^∞ controllers. Thus, in this case, Theorem 1 is reduced to the results of the standard H^∞ control problems [2], [3].

The rather implicit specification of all solutions in Theorem 1 will be replaced by the conditions involving interpolation constraints on $S(s)$ in next section.

4 Interpolation Conditions

In order to find $S(s) \in BH^\infty$ such that $K(s)$ in (44) is proper and stabilizes the closed-loop system, we apply the well-known Youla parameterization to $P_n(s)$ as:

LEMMA 4 Every $K(s)$ which internally stabilizes $P_n(s)$ is given by

$$K(s) = F_l(M(s), Q(s)), \quad (50)$$

where $Q(s) \in RH^\infty$ is a free parameter, and

$$M(s) = \left[\begin{array}{cc|cc} A_n + B_{n2}F + HC_{n2} & -H & B_{n2} & \\ \hline F & 0 & I_p & \\ C_{n2} & -I_q & 0 & \end{array} \right]. \quad (51)$$

with F and H being any matrices which stabilizes $A_{nF} := A_n + B_{n2}F$ and $A_{nH} := A_n + HC_{n2}$, respectively.

Then H^∞ control problem for plant $P_n(s)$ can be reduced to the following 1-block plant model matching problem: Find $Q(s) \in RH^{p \times q}$ such that

$$\|\Phi_n\|_\infty < 1, \quad \Phi_n(s) := T_{n1}(s) + T_{n2}(s)Q(s)T_{n3}(s) \quad (52)$$

holds, where

$$P_Q = \left[\begin{array}{cc} T_{n1} & T_{n2} \\ T_{n3} & 0 \end{array} \right] := \left[\begin{array}{cc|cc} A_{nF} & -B_{n2}F & B_{n1} & B_{n2} \\ \hline 0 & A_{nH} & B_{nH} & 0 \\ C_{nF} & -D_{12}F & 0 & N_{12} \\ 0 & -C_{n2} & -N_{21} & 0 \end{array} \right]. \quad (53)$$

where $C_{nF} := C_1 + N_{12}F$, $B_{nH} := B_{n1} + HN_{21}$.

Since $P_{n12}(s)$ and $P_{n21}(s)$ do not have invariant zeros in C_+ , neither does $T_{n1}(s)$ nor $T_{n2}(s)$. We have

THEOREM 2 $S(s) \in BH^\infty$ such that $K(s)$ in (44) is proper and stabilizes the closed-loop system if and only if $S(s) \in BH^\infty$ such that

$$Q = T_{n2}^{-1}(S - T_{n1})T_{n3}^{-1} \in RH^\infty \quad (54)$$

holds.

Proof Necessary Suppose $S(s) \in BH^\infty$ such that $K(s)$ in (44) is proper and stabilizes the closed-loop system, then $F_l(P_n, K) = S(s)$. From (50) and (52), since there is one to one transformation relation between $Q(s)$ and $K(s)$, and $\Phi_n(s)$ is invariant, then $\Phi_n(s) = F_l(P_n, K) = S(s)$. Thus (54) holds.

Sufficiency Suppose that there exists $S(s) \in BH^\infty$ such that (54) holds. Rewrite (54) as

$$S = F_l(P_Q, Q), \quad (55)$$

where P_Q is given in (53). Put the above equation into (44), we obtain

$$K(s) = F_l(M^\infty, S) = F_l(M^\infty * P_Q, Q) = F_l(M, Q).$$

where $M^\infty * P_Q = M$ can be derived by verifying

$$P_Q = J(M^\infty)^{-1}J * M. \quad (56)$$

Since $(M^\infty)^{-1}$ is proper from (45), direct calculation shows that (56) holds. Therefore, $K(s)$ in (44) is proper and stabilizes the closed-loop system. \square

For the simplicity of the notation, we will drop the subscript n in T_{ni} ($i = 1, 2, 3$) in the rest of this section.

The following steps are taken to find the constraints on $S(s)$ according to Theorem 2, where Steps 1 and 2 are according to [4].

Step 1 Get the Smith form of $T_i(s)$ ($i = 2, 3$) over RH^∞ as

$$T_i(s) = U_{iL}(s)\Lambda_i(s)U_{iR}(s) = U_{iL}(s)V_i(s)W_i(s)U_{iR}(s) \quad (57)$$

where $U_{iL}(s)$ and $U_{iR}(s)$, $i = 2, 3$, be unimodular matrices, with $\Lambda_i(s) \in RH^\infty$, and with V_i, W_i diagonal matrices such that

$$\Lambda_i = V_iW_i, \quad (58)$$

where $V_i \in RH^\infty$ possesses zeros in C_- and $W_i \in RH^\infty$ have all zeros on Ω_e .

Step 2 Let

$$\hat{T}_2(s) = U_{2L}(s)V_2(s) \quad \hat{T}_3(s) = V_3(s)U_{3R}(s) \quad (59)$$

Then $\hat{T}_2(s)$ and $\hat{T}_3(s)$ are unimodular in RH^∞ . Get

$$Q = U_{2R}^{-1}W_2^{-1}\hat{T}_2^{-1}(S - T_1)\hat{T}_3^{-1}W_3^{-1}U_{3L} \quad (60)$$

Note that the matrices W_2, W_3 have zeros on Ω_e . With U_{2R}, U_{3L} unimodular over RH^∞ and $Q \in RH^\infty$ it is necessary that the zeros on Ω_e of W_2, W_3 are those of $\hat{T}_2^{-1}(S - T_1)\hat{T}_3^{-1}$

Step 3 Suppose in particular that z_2 is a zero of the i th diagonal entry of W_2 of order j , but is not a zero of any entry of W_3 . Let e_i be a column vector with its i th entry being 1 and remainders being 0. Using the unimodularity over RH^∞ of \hat{T}_3^{-1} , it follows that

$$\frac{d^k}{ds^k} e_i^T \left[\hat{T}_2^{-1}(S - T_1) \right] \Big|_{z_2} = 0 \quad k = 0, 1, \dots, j-1 \quad (61)$$

which is of the form

$$\begin{aligned} \alpha_{i0} S(z_2) + \beta_{i0} &= 0 \\ \alpha_{i0} S'(z_2) + \alpha_{i1} S(z_2) + \beta_{i1} &= 0 \\ \alpha_{i0} S''(z_2) + 2\alpha_{i1} S'(z_2) + \alpha_{i2} S(z_2) + \beta_{i2} &= 0 \\ &\vdots \end{aligned} \quad (62)$$

for complex vectors $\alpha_{i0}, \beta_{i0}, \alpha_{i1}, \beta_{i1}, \dots$

Step 4 Similarly, if z_3 is zero of the p th diagonal entry of W_3 of order q , but is not a zero of any entry of W_2 ,

$$\frac{d^k}{ds^k} \left[(S - T_1) \hat{T}_3^{-1} \right] e_p \Big|_{z_3} = 0 \quad k = 0, 1, \dots, q-1 \quad (63)$$

This leads to right hand interpolation conditions of the form

$$\begin{aligned} S(z_3)\gamma_{p0} + \delta_{p0} &= 0 \\ S'(z_3)\gamma_{p0} + S(z_3)\gamma_{p1} + \delta_{p1} &= 0 \\ S''(z_3)\gamma_{p0} + 2S'(z_3)\gamma_{p1} + S(z_3)\gamma_{p2} + \delta_{p2} &= 0 \\ &\vdots \end{aligned} \quad (64)$$

Step 5 If z_1 is a j th order zero of the i th diagonal entry of W_2 and a q th order zero of the p th diagonal entry of W_3 , then (i, p) entry of $\hat{T}_2^{-1}(S - T_1)\hat{T}_3^{-1}$ must have zero z_1 of order $j + q$. Therefore, in addition to (61) and (63), we have

$$\frac{d^k}{ds^k} \left[e_i^T \hat{T}_2^{-1}(S - T_1)\hat{T}_3^{-1} e_p \right] \Big|_{z_1} = 0 \quad k = 0, 1, \dots, j+q-1. \quad (65)$$

The interpolation conditions are not quite as tidy as in the previous two cases. However, they are perhaps not as complicated as may at first seem the case. If $j = q = 1$, then the fact that z_1 is a zero of the i th diagonal entry of W_2 implies

$$e_i^T [\hat{T}_2^{-1}(z_1)(S(z_1) - T_1(z_1))] = 0 \quad (66)$$

The fact that z_1 is a zero of the p th diagonal entry of W_3 implies

$$(S(z_1) - T_1(z_1))T_3^{-1}(z_1)e_p = 0 \quad (67)$$

and the fact that it is a *double* zero of the (i, p) entry of $\hat{T}_2^{-1}(S - T_1)\hat{T}_3^{-1}$ yields additionally

$$\frac{d}{ds} \left[e_i^T T_2^{-1}(S - T_1)\hat{T}_3^{-1} e_p \right] \Big|_{z_1} = 0$$

The above calculation is not much more complicated in case one of j or q is greater than 1.

We have now proved the "only if" part of the following Theorem.

THEOREM 3 $S(s) \in BH^\infty$ such that $K(s)$ in (44) is proper and stabilizes the closed-loop system if and only if $S(s) \in BH^\infty$ satisfies interpolation conditions of the type depicted in (61), (63), and (65).

Proof ("If" statement). Suppose that the interpolation constraints hold. Then it is not difficult to see that

necessarily $\hat{T}_2^{-1}(S - T_1)\hat{T}_3^{-1} = W_2 R W_3$ for some $R \in RH_{p \times q}^\infty$. It follows that $Q = U_{2R}^{-1} R U_{3L}^{-1} \in RH^{p \times q}$, since U_{2R}, U_{3L} are unimodular. This completes the proof with Theorem 2.

In summary of the discussion in this paper, suppose that the H^∞ control problem is solvable, we give the following algorithm for calculation of all H^∞ controllers.

Algorithm

1. Get $P_n(s)$ in (35) via solving two Riccati equations in Section 2. And apply the Youla parameterization to $P_n(s)$ to get $T_i(s)$, $i = 1, 2, 3$.
2. Get the Smith form of $T_i(s)$ over RH^∞ to get $\hat{T}_i(s)$ and W_i , $i = 2, 3$.
3. Find all $S(s) \in BH^\infty$ satisfy interpolation conditions of the type depicted in (61), (63) and (65).
4. Get $K(s)$ in (44) with the above calculated $S(s) \in BH^\infty$.

5 Conclusion

In this paper, we have described a parametrization of all matrices of the 4-block H^∞ control problem with infinite and finite $j\omega$ -axis zeros. The parametrization is in terms of transfer function matrices in $BH_{p \times q}^\infty$ satisfying interpolation constraints.

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